

Outage Probability for Joint Processing Coordinated Multi-Point (JP-CoMP) Performance Analysis in LTE-Advanced Networks

Marceau Coupechoux*

* TELECOM ParisTech, INFRES/RMS

joint work with D. Ben Cheikh (Orange Labs/TPT), J.M. Kelif (Orange Labs)
and Ph. Godlewski (TPT)

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Cellular Networks Evolution (3GPP) I

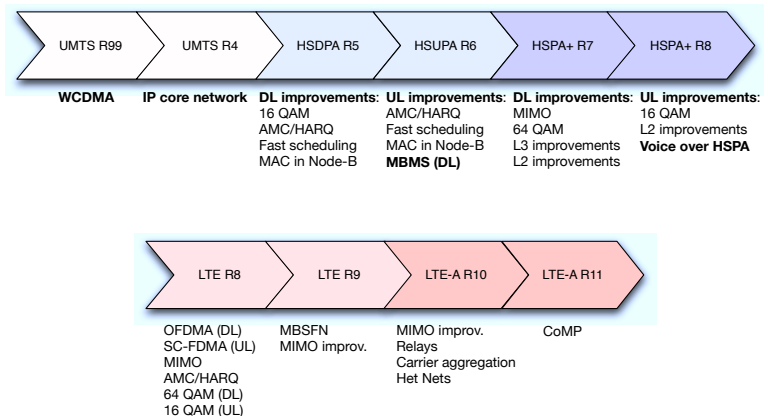
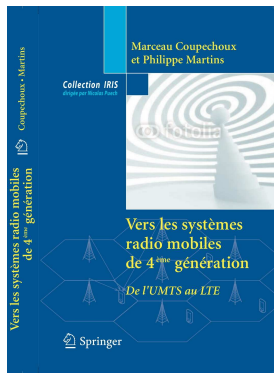


Figure: Evolution of the 3GPP Standards.

Cellular Networks Evolution (3GPP) II



M. Coupechoux and Ph. Martins, "Vers la quatrième génération" (Towards Fourth Generation), Springer Verlag, 315 pp, to appear (hopefully) mid 2012.

LTE

• LTE Requirements:

- Peak data rates: 300 (DL) / 75 Mbps (UL)
- Peak spectrum efficiency: 15 (DL) / 3.75 bps/Hz (UL)
- Average spectrum efficiency: 2.67 (4x4 DL) / 0.74 (UL 1x2) bps/Hz/cell
- U-plane latency < 5 ms
- Spectrum flexibility: from 1.4 to 20 MHz

• LTE Key Features:

- OFDMA (DL) / SC-FDMA (UL)
- Adaptive Modulation (up to 64-QAM) and Coding / HARQ
- MIMO (up to 4x4): Spatial Multiplexing, Transmit Diversity, Beamforming, MU-MIMO
- Frequency selective fast scheduling
- IP-based flat architecture

LTE-Advanced Release 10

- LTE-Advanced Requirements:
 - Peak data rates: 1 Gbps (DL) / 500 Mbps (UL)
 - Peak spectrum efficiency: 30 (DL) / 15 bps/Hz (UL)
 - Average spectrum efficiency: 3.7 (4x4 DL) / 2.0 (UL 2x4) bps/Hz/cell
- LTE-Advanced Rel. 10 Features:
 - Carrier aggregation (up to 100 MHz)
 - Clustered SC-FDMA (UL)
 - MIMO: 8x8 (DL), 4x8 (UL), enhanced MU-MIMO
 - Relaying
 - Heterogeneous Networks (HetNets)

Coordinated Multi-Point for LTE-A Rel. 11

- **Goal:** coordinate BS transmissions in order to reduce interference and improve cell-edge throughput.
- **CS/CB:** scheduling and beamforming are coordinated among clustered BS.
- **JP:** data is transmitted from several BS to the same UE *coherently* (tight synchronization is needed) or *non-coherently* (gain obtained from power boost).

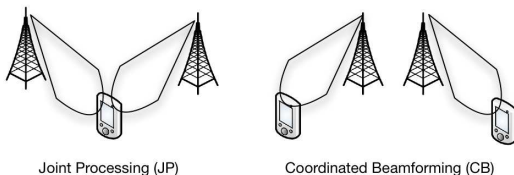


Figure: Coordinated Multi-Point (CoMP).

Outage Probability and SIR CDF

SINR/SIR CDF is a crucial input for cellular network dimensioning:

- **Coverage studies:**

$P[\gamma < \gamma_{th}]$, where γ_{th} is the the common control channel SIR threshold provides coverage; γ_{th} s.t.

$P[\gamma < \gamma_{th}] = 10\%$ provides cell-edge throughput.

- **Capacity studies:**

channel state distribution
+ M/G/1-PS (or other Markov chain [25]) approach provides capacity figures.

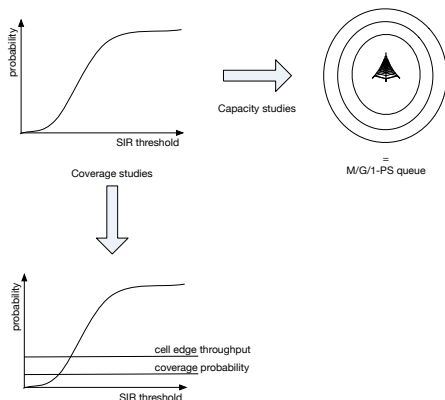


Figure: Outage probability and dimensioning issues.

Outlines

- Related Work
- Maximum Ratio Transmission
- System Model
- Outage Probability
- Numerical Results

Related Work I

- Performance of CoMP has been asserted in several works:
 - The JP strategy offers larger performance gain than the CS/CB [1, 2].
 - Multicell cooperation attain larger capacity than an isolated cell (measurement study [3]).
 - A field trial [4] confirms the throughput enhancement introduced by JP-CoMP.
 - In [5], different JP schemes are proposed and compared numerically.
 - In [6], zero forcing (ZF) and maximum ratio transmission (MRT) schemes are studied for femto cell coordination. Two power allocation algorithms are proposed and compared.
 - In [3, 4, 5, 6], authors performed simulation, measurement or field study but **no theoretical studies** were conducted.

Related Work II

- There are few analytical works:
 - In [7], an analytical expression of the capacity outage probability was derived for an open-loop Alamouti-like CoMP downlink transmission in Rayleigh fading. **The proposed SINR expression can only be achieved when using a distributed Alamouti for two cooperating BSs.**
 - In [8], an analytical study of a multi-cell multi-antenna cooperative MRT/MRC scheme was conducted. An analytical expression of the PDF of the SIR was derived considering path-loss, shadowing and Rayleigh fading. **Limitations of the study** are:
 - User is supposed to be at equal distances from cooperative BSs,
 - Shadowing is modeled with a Gamma distribution,
 - Transmitters are supposed to be Poisson distributed,
 - Significant differences bw. simulations and theoretical results.
- Several works on feedback constraints.
- Several works on BS selection algorithms.

Maximum Ratio Transmission I

Assume n_t Tx antennas and n_r Rx antennas without interference, then the received signal is:

$$y = Hws + n,$$

where

- H is the $n_r \times n_t$ channel gain matrix,
- w is the $n_t \times 1$ precoding vector with $\|w\|^2 \leq 1$,
- s is the information symbol, n is AWGN.

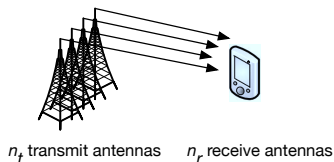


Figure: MRT Model.

Maximum Ratio Transmission II

This equation can be interpreted as an AWGN channel with SNR:

$$\gamma = \frac{||Hw||^2}{\sigma_n^2} E[|s|^2].$$

The scheme that maximizes $||Hw||^2$ under the constraint $||w||^2 \leq 1$ is called MRT.

In general:

$$\frac{||Hw||^2}{||w||^2} \leq \lambda_{\max}(H^H H),$$

and the optimal precoder is the normalized eigenvector associated to the largest eigenvalue of $H^H H$.

Maximum Ratio Transmission III

If $n_r = 1$, then $H = h$ and using Cauchy-Schwartz inequality:

$$|hw|^2 \leq ||h||^2 ||w||^2 = ||h||^2,$$

with equality if w is proportional to h^* .

The **MRT precoder** can thus be written:

$$w_{MRT} = \frac{h^*}{||h||}.$$

MRT maximizes the received signal power.

Drawbacks of the scheme:

- Interference is not taken into account.
- Perfect channel information is required at transmitter.

System Model I

We consider the following system model:

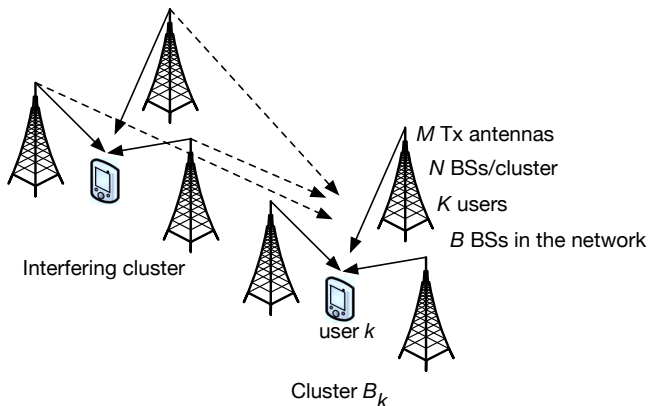


Figure: System Model.

System Model II

We assume that JP-CoMP is based on MRT [14].

Received signal can be written:

$$y_k = \sum_{b \in B_k} \sqrt{p_{b,k}} \mathbf{h}_{b,k} \mathbf{x}_{b,k} + \sum_{i=1, i \neq k}^K \sum_{j \in B_i} \sqrt{p_{j,k}} \mathbf{h}_{j,k} \mathbf{x}_{j,i} + n, \quad (1)$$

where

- $p_{b,k}$ is the power received by user k from BS b , n is the AWGN,
- $\mathbf{h}_{b,k} \in \mathbb{C}^{1 \times M}$ is the complex Gaussian channel bw. k and b ,
- $\mathbf{x}_{b,k} \in \mathbb{C}^{M \times 1}$ is the MRT data vector:

$$\mathbf{x}_{b,k} = \frac{\mathbf{h}_{b,k}^*}{\|\mathbf{h}_{b,k}\|} s_k. \quad (2)$$

System Model III

The received power $p_{b,k}$ includes path-loss and shadowing:

$$p_{b,k} = P_{Tx} C r_{b,k}^{-\eta} 10^{\frac{\xi_{b,k}}{10}}, \quad (3)$$

where P_{Tx} is the total transmit power of BS b , C is a constant, $r_{b,k}$ is the distance bw. k and b , η is the path-loss exponent and $\xi_{b,k}$ is $\mathcal{N}(0, \sigma)$. Shadowing is now assumed to be constant.

The output SIR perceived by a user k is given by:

$$\gamma_k = \frac{\left(\sum_{b \in B_k} \sqrt{p_{b,k}} \|\mathbf{h}_{b,k}\| \right)^2}{\sum_{i=1, i \neq k}^K \left| \sum_{j \in B_i} \sqrt{p_{j,k}} \mathbf{h}_{j,k} \frac{\mathbf{h}_{j,i}^*}{\|\mathbf{h}_{j,i}\|} \right|^2} = \frac{X}{Y}. \quad (4)$$

Outage Probability I

Useful signal CDF

We can write:

$$X = U^2, \quad U = \sum_{b \in B_k} \sqrt{p_{b,k}} \|\mathbf{h}_{b,k}\|. \quad (5)$$

U can also be written as:

$$U = \sum_{b \in B_k} \sqrt{p_{b,k}} \sqrt{\sum_{i=1}^M |h_{b,k,i}|^2}. \quad (6)$$

Closed form expression of the PDF of U is not known, we thus rely on the central limit theorem for causal functions.

Outage Probability II

Theorem (CLCF)

Consider n RV $X_i \geq 0$ with density f_i , means η_i and variances σ_i^2 . Let $X = X_1 + \dots + X_n$ with density $f(t) = f_1(t) \star f_2(t) \dots \star f_n(t)$. Denote

$$\begin{aligned}\eta &= \eta_1 + \eta_2 + \dots + \eta_n, \\ \sigma^2 &= \sigma_1^2 + \sigma_2^2 + \dots + \sigma_n^2.\end{aligned}$$

Under certain conditions, $f(t) \sim \frac{t^\alpha e^{-t/\beta}}{\beta^{\alpha+1} \Gamma(\alpha+1)}$ as $n \rightarrow \infty$,
where $\alpha + 1 = \frac{\eta^2}{\sigma^2}$ and $\beta = \frac{\sigma^2}{\eta}$.

Proof.

see [13].



Outage Probability III

The use of the CLCF provides:

$$f_U(u) = \frac{u^{\nu-1} e^{-\frac{u}{\theta}}}{\Gamma(\nu)\theta^\nu}. \quad (7)$$

where $\nu = \frac{E[U]^2}{\text{var}(U)}$ and $\theta = \frac{\text{var}(U)}{E[U]}$. Denoting $V = \sqrt{\sum_{i=1}^M |h_{b,k,i}|^2}$, V is a square root of a Gamma RV and thus we can write the PDF of V as:

$$f_V(v) = \frac{2}{(M-1)!} v^{2M-1} e^{-v^2}, \quad (8)$$

$E[V]$ can be derived using (8) and is given by:

$$E[V] = \int_0^\infty \frac{2}{(M-1)!} v^{2M} e^{-v^2} dv = \frac{(2M-1)!!}{2^M(M-1)!} \sqrt{\pi}, \quad (9)$$

Outage Probability IV

So that:

$$\mathbb{E}[U] = \frac{(2M-1)!!}{2^M(M-1)!} \sqrt{\pi} \sum_{b \in B_k} \sqrt{p_{b,k}}. \quad (10)$$

On the other hand:

$$\begin{aligned} \text{var}(U) &= \sum_{b \in B_k} p_{b,k} \mathbb{E}[\|\mathbf{h}_{b,k}\|^2] - \sum_{b \in B_k} p_{b,k} \mathbb{E}[\|\mathbf{h}_{b,k}\|]^2 \\ &= (M - \pi(\frac{(2M-1)!!}{2^M(M-1)!})^2) \sum_{b \in B_k} p_{b,k}. \end{aligned} \quad (11)$$

Note: $(2N)!! = 2 \times 4 \times \cdots \times (2N)$ and $(2N+1)!! = 3 \times 5 \times \cdots \times (2N+1)$.

Outage Probability V

The parameters ν and θ are thus given by:

$$\nu = \frac{(2M-1)!!^2}{M2^{2M}(M-1)!^2 - \pi(2M-1)!!^2} \frac{(\sum_{b \in B_k} \sqrt{p_{b,k}})^2}{\sum_{b \in B_k} p_{b,k}}, \quad (12)$$

$$\theta = \frac{M2^{2M}(M-1)!^2 - \pi(2M-1)!!^2}{2^M(M-1)!(2M-1)!!\sqrt{\pi}} \frac{\sum_{b \in B_k} p_{b,k}}{\sum_{b \in B_k} \sqrt{p_{b,k}}}. \quad (13)$$

From (10) and (11):

- Mean useful signal power increases with N .
- Variability also increases with N .
- It can be shown that mean and variance increase with M .

Outage Probability VI

Using the Gamma approximation of the PDF of U , we can derive the CDF of X as follows:

$$F_X(v) = \int_0^{\sqrt{v}} \frac{u^{\nu-1} e^{-\frac{u}{\theta}}}{\Gamma(\nu)\theta^\nu} du, \quad (14)$$

$$= \frac{\gamma(\nu, \frac{\sqrt{v}}{\theta})}{\Gamma(\nu)}, \quad (15)$$

where $\gamma(.,.)$ is the lower incomplete gamma function:

$$\gamma(a, x) = \int_0^x t^{a-1} e^{-t} dt.$$

Outage Probability VII

Interference PDF

We can write now:

$$Y = \sum_{i=1, i \neq k}^K \left| \sum_{j \in B_i} \sqrt{p_{j,k}} g_{j,k,i} \right|^2, \quad (16)$$

where $g_{j,k,i} = \mathbf{h}_{j,k} \frac{\mathbf{h}_{j,i}^*}{\|\mathbf{h}_{j,i}\|}$.

$g_{j,k,i}$ is complex Gaussian independent of $\mathbf{h}_{j,i}$.

Outage Probability VIII

Claim ([15])

Consider zero-mean complex Gaussian vectors $\mathbf{h}_0 = [h_{1,0}, h_{2,0}, \dots, h_{N,0}]^H$ and $\mathbf{h}_j = [h_{1,j}, h_{2,j}, \dots, h_{N,j}]^H$. Let $g_j = \frac{\mathbf{h}_0^H \mathbf{h}_j}{\|\mathbf{h}_0\|}$. Then, g_j is independent on \mathbf{h}_0 .

Proof.

$g_j | \mathbf{h}_0$ is zero-mean complex Gaussian and:

$$E[g_j | \mathbf{h}_0] = \frac{\mathbf{h}_0^H}{\|\mathbf{h}_0\|} E[\mathbf{h}_j] = 0$$

$$E[|g_j|^2 | \mathbf{h}_0] = \frac{\mathbf{h}_0^H E[\mathbf{h}_j \mathbf{h}_j^H] \mathbf{h}_0}{\|\mathbf{h}_0\|^2} = \frac{\mathbf{h}_0^H I_N \mathbf{h}_0}{\|\mathbf{h}_0\|^2} = 1$$



Outage Probability IX

Let $c_{k,i} = \sum_{j \in B_i} \sqrt{p_{j,k}} g_{j,k,i}$, it is the sum of independent complex Gaussian random variables. Y can be written as:

$$Y = \sum_{i=1, i \neq k}^K |c_{k,i}|^2, \quad (17)$$

$\{c_{k,i}\}_{(i=1 \dots K, i \neq k)}$ being independent zero-mean complex Gaussian elements with variances:

$$\lambda_{i,k} = \text{var}(c_{k,i}) = \sum_{j \in B_i} p_{j,k}. \quad (18)$$

Outage Probability X

The PDF of Y is, hence, given by [16]:

$$f_Y(y) = \sum_{i=1, i \neq k}^K \frac{\Pi_i}{\lambda_{i,k}} \exp\left(-\frac{y}{\lambda_{i,k}}\right), \quad (19)$$

where $\Pi_i = \prod_{p=1..K, p \neq i, p \neq k} \frac{\lambda_{p,k}}{\lambda_{p,k} - \lambda_{i,k}}$.

Having the PDF of X and the PDF of Y , and since they are independent RV, we derive the outage probability as follows:

$$P(\gamma_k < \gamma_{th}) = \int F_X(\gamma_{th}y) f_Y(y) dy, \quad (20)$$

$$= \sum_{i=1, i \neq k}^K \frac{\Pi_i}{\lambda_{i,k} \Gamma(\nu)} \int \exp\left(-\frac{y}{\lambda_i}\right) \gamma\left(\nu, \frac{\sqrt{\gamma_{th}y}}{\theta}\right) dy. \quad (21)$$

Outage Probability XI

The outage probability is given by [17] [18]:

$$P(\gamma_k < \gamma_{th}) = \sum_{i=1, i \neq k}^K \Pi_i \left(\frac{\sqrt{\gamma_{th} \lambda_{i,k}}}{2\theta} \right)^\nu U\left(\frac{\nu}{2}, \frac{1}{2}, \frac{\gamma_{th} \lambda_{i,k}}{4\theta^2}\right) \text{ (CoMP)}$$

where $U(.,.,.)$ is the confluent hypergeometric function of second kind:

$$U(a, b, z) = \frac{1}{\Gamma(a)} \int_0^\infty e^{-zt} t^{a-1} (1+t)^{b-a-1} dt.$$

Numerical Results I

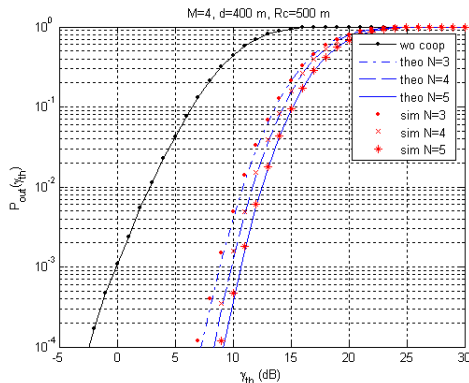


Figure: Comparison between simulated and analytical results of the outage probability without CoMP and with CoMP MRT strategy for 3, 4 or 5 cooperating BSs.

Numerical Results II

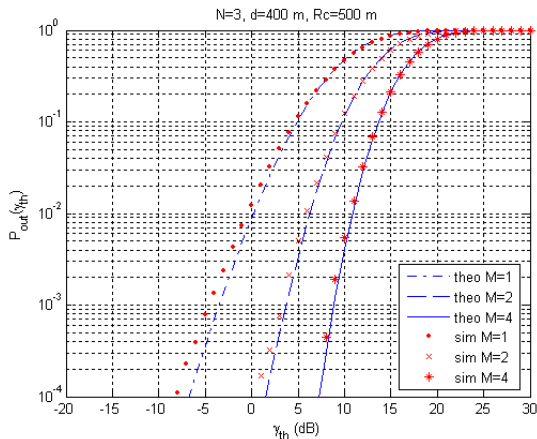


Figure: Influence of the number of transmit antennas.

Numerical Results III

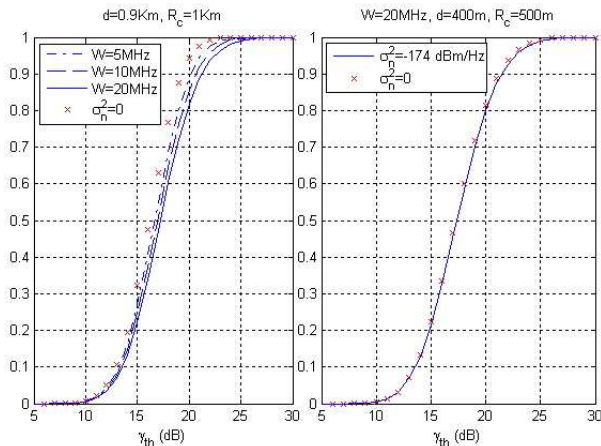


Figure: Impact of the noise power on the outage probability.

Numerical Results IV

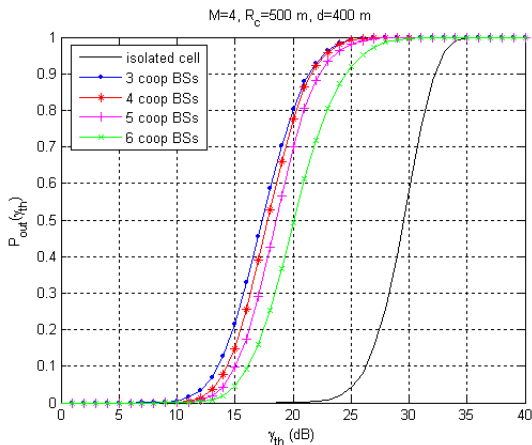


Figure: Comparison with an isolated cell.

Other Obtained Results I

OFDMA Cellular networks [19]:

$$\mathbb{P}(SIR_{eff} < \delta) = 1 - Q \left[\frac{\log_2(1 + \delta) - \mu_{MIC}}{\sigma_{MIC}} \right]. \quad (\text{OFDMA})$$

$$\mu_{MIC} = E[C_n], \quad (22)$$

$$\sigma_{MIC}^2 = \frac{1}{N_{sc}} (E[C_n^2] - E[C_n]^2). \quad (23)$$

$$E[C_n] = \int \int (1 - Q_f(x, t)) e^{-x} dx dt, \quad (24)$$

$$E[C_n^2] = \int \int 2t (1 - Q_f(x, t)) e^{-x} dx dt. \quad (25)$$

$$Q_f(x, t) = Q \left[\frac{10 \log_{10}(\frac{x}{2^t - 1}) - m_f}{s_f} \right]. \quad (26)$$

Other Obtained Results II

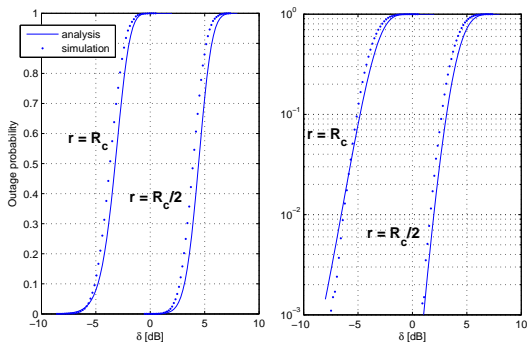


Figure: Outage probability at cell edge ($r = R_c$) and inside the cell ($r = R_c/2$); comparison between analysis (solid curves) and simulations (dotted curves) on a hexagonal network ($N_{sc} = 48$, $\eta = 3$, $\sigma = 6$ dB).

Other Obtained Results III

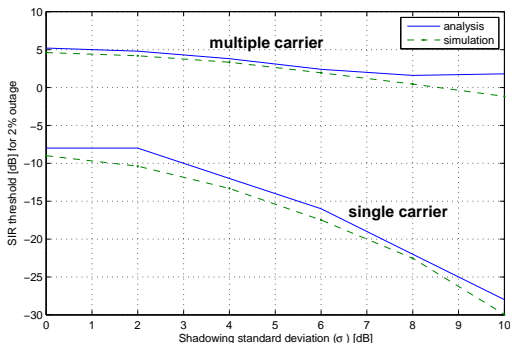


Figure: SIR threshold δ in the cell ($r = R_c/2$) for 2% outage probability as a function of the shadowing standard deviation; comparison between analysis (solid curves) and simulations (dotted curves) on a hexagonal network ($N_{sc} = 48$, $\eta = 3$).

Other Obtained Results IV

MIMO Alamouti transmission [20]:

$$P_{out}(\gamma_{th}) = 1 - \left(\frac{p_0}{2\gamma_{th}\beta + p_0} \right)^\alpha \left(1 + \frac{2\gamma_{th}\beta}{2\gamma_{th}\beta + p_0} \frac{\Gamma(\alpha+1)}{\Gamma(\alpha)} \right). \text{ (ALAM)}$$

$$\alpha = \frac{2}{1 + \rho_z} \frac{(\sum_{j=1}^B p_j)^2}{\sum_{j=1}^B p_j^2}, \quad (27)$$

$$\beta = \frac{1 + \rho_z}{2} \frac{\sum_{j=1}^B p_j^2}{\sum_{j=1}^B p_j}. \quad (28)$$

$$\rho_z = 0.0167 \quad (29)$$

Other Obtained Results V

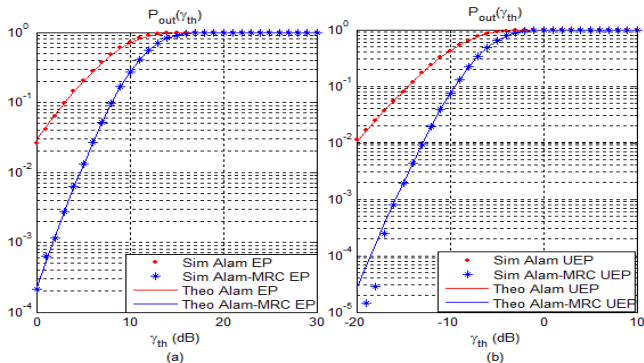


Figure: P_{out} versus SINR threshold for 2×1 MISO Alamouti and 2×2 MIMO Alamouti with MRC receiver systems in case of equal (a) and unequal (b) received interference power (EP and UEP resp. in legends).

Other Obtained Results VI

MU-MIMO Zero-Forcing/Beamforming [21]:

$$P_{out}(\gamma_{th}) = \frac{\Gamma(\nu + \alpha)}{\Gamma(\nu)\mu^\nu\alpha!} \frac{p_0^\nu (K\gamma_{th})^\alpha}{(\gamma_{th}K + \frac{p_0}{\mu})^{\nu+\alpha}} {}_2F_1(1, \nu + \alpha; \alpha + 1; \frac{\gamma_{th}K}{\gamma_{th}K + \frac{p_0}{\mu}})$$

(ZF)

where ${}_2F_1(., .; .; .)$ is the Gauss' hypergeometric function [17],

$$\nu = \left(1 - \frac{2(2\alpha - 1)\theta^2 - (\alpha - 1)^2(\alpha - 2)}{(5\alpha - 4)\theta^2 + 2(\alpha - 1)(\alpha - 2)\theta}\right) \frac{\left(\sum_{j=1}^B p_j\right)^2}{\sum_{j=1}^B p_j^2} \quad (30)$$

$$\mu = \frac{(5\alpha - 4)\theta^2 + 2(\alpha - 1)(\alpha - 2)\theta}{(\alpha - 1)(\alpha - 2)\theta + (\alpha - 1)^2(\alpha - 2)} \frac{\sum_{j=1}^B p_j^2}{\sum_{j=1}^B p_j} \quad (31)$$

$$\alpha = M - K + 1, \quad \theta = (1 - \rho^2), \quad \rho = 0.154 \quad (32)$$

Other Obtained Results VII

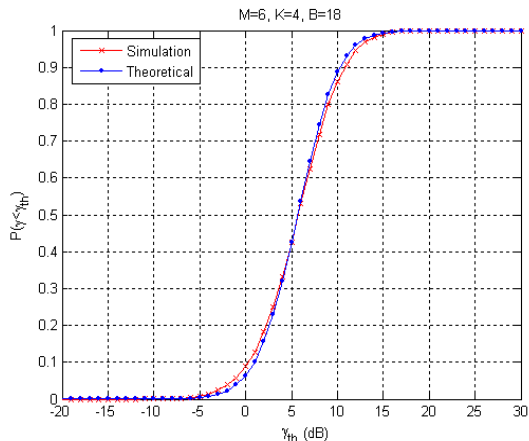


Figure: Outage probability of a zero forcing multicellular precoded system (4 users, 6 antennas at the BS).

Conclusion

We have derived an approximate closed form expression of the SIR CDF.
Further work:

- Consider more realistic scenarios
- Compare with other beamforming schemes
- Consider dynamic studies

Outage probabilities can be efficiently used for solving various dimensioning issues:

- Effect of network densification [22]
- Limiting transmit power in green cellular networks [23]
- Perf. evaluation of frequency reuse schemes in OFDMA networks [24]
- Dynamic performance study of WiMAX networks [25]

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